

## Waves at Higher Frequency - Dispersion

$$\nabla \cdot \vec{E} = \rho_{\text{mat}}$$

$$\nabla \times \vec{B} = \frac{\vec{j}_{\text{mat}}}{c} + \frac{1}{c} \partial_t \vec{E}$$

$$\nabla \cdot \vec{B} = 0$$

$$\nabla \times \vec{E} = -\frac{1}{c} \partial_t \vec{B}$$

• Generally have been assuming  $\omega \ll \frac{1}{\tau_{\text{micro}}}$

$$k \ll \frac{1}{l_{\text{micro}}} \quad \text{or} \quad \lambda \gg l_{\text{micro}}$$

Certainly this is far from clear in the optical range

$$h\omega = hc \frac{\omega}{c} = hc \frac{2\pi}{\lambda}$$

$$= 197 \text{ eV} \cdot \text{nm} \cdot \frac{2\pi}{600 \text{ nm}}$$

) for  $\lambda = 600 \text{ nm}$

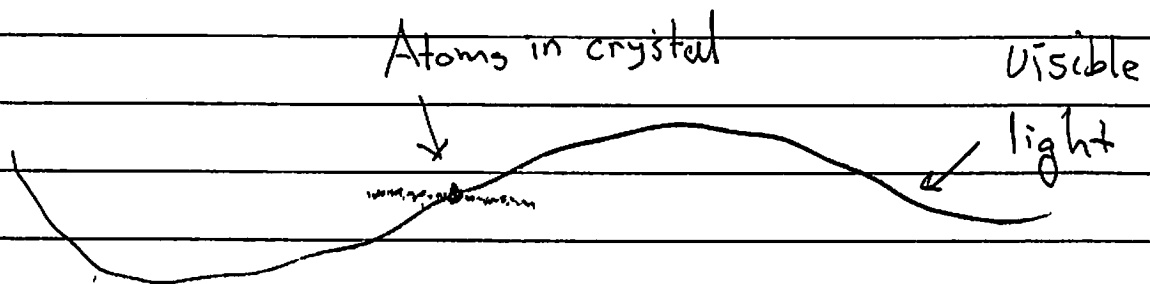
$h\omega = 2.0 \text{ eV}$  of order atomic energies

However, note

$$\lambda \sim 600 \text{ nm} \sim 6000 \text{ \AA}$$

That  $\lambda \gg$  atomic sizes  $\sim 0.5 \text{ \AA}$

So we can still expand the current in spatial gradients but need



to consider the atomic response times.

$$\nabla \cdot E = \rho_{\text{mat}}(t)$$

$$\nabla \times B = j_{\text{mat}}(t)/c + \frac{1}{c} \partial_t E$$

$$\nabla \cdot B = 0$$

$$\nabla \times E = -\frac{1}{c} \partial_t B$$

What is  $j_{\text{mat}}$ ?

## Linear Response for $\vec{j}_{\text{mat}}$

In general:

$\vec{j}(t, x)$  Depends on the past values of the fields in a linear approximation

The most general linear form involving no spatial derivatives that is allowed by parity

$$\vec{j}(t) = \int dt' \underbrace{\sigma(t-t')}_{\text{response function}} \vec{E}(t')$$

Clearly for a causal system  $\vec{j}(t)$  depends on  $\vec{E}(t')$  for  $t' < t$ . Thus we have

$$\sigma(t) = 0 \quad \text{for } t < 0 \quad (\text{i.e. } t' > t)$$

Then in frequency space

$$\vec{j}(\omega) = \sigma(\omega) \vec{E}(\omega)$$

↖ frequency dependent conductivity

## Linear Response pg. 2

Can continue and add the first derivatives:

$$j_{\text{mat}}(\omega) = -i\omega \chi_e(\omega) \vec{E}(\omega) + c \chi_m^B(\omega) \nabla \times B(\omega)$$

Then from current conservation

$$\partial_t \rho + \nabla \cdot j = 0 \quad \Leftrightarrow \quad \rho(\omega) = \nabla \cdot j(\omega) / (-i\omega)$$

we have since  $\nabla \cdot (\nabla \times B(\omega)) = 0$ ,

$$\rho_{\text{mat}}(\omega) = -\chi_e(\omega) \nabla \cdot E$$

Thus the only difference from before is now  $\chi_e(\omega)$  and  $\chi_m^B(\omega)$  are functions of  $\omega$ . Always complex functions

$$E(\omega) \nabla \cdot E = 0$$

$$\nabla \times B = \frac{\epsilon(\omega) \mu(\omega)}{c^2} (-i\omega E)$$

$$\nabla \cdot B = 0$$

$$\nabla \times E = \frac{+i\omega B}{c}$$

where (as before)

$$\epsilon(\omega) = 1 + \chi_e(\omega) \quad \text{and} \quad \mu(\omega) = \frac{1}{1 - \chi_m(\omega)}$$

## Last Time

- Reflection of waves at interfaces
  - stress
- Waves in metals, skin depth
- Reflection of waves in metals

## Today

- Discuss waves at higher frequency,  $T \sim T_{\text{micro}}$ ,  
but still long wavelength,  $L \gg l_{\text{micro}}$ .

period of wave  
↓  
atomic time scales  
↓

$$c \sim \frac{L}{T} \gg \frac{l_{\text{micro}}}{T_{\text{micro}}} \sim v_{\text{micro}} \sim \alpha c$$

electrons in atoms moving non-relativistic  
↑  
 $\frac{1}{137}$

- Will argue first that for linear media, and for waves of a definite frequency  $\omega$  can just consider  $\epsilon(\omega)$ , and  $\mu(\omega)$ , effective frequency dependent dielectric constants:

$$\epsilon, \mu \longrightarrow \hat{\epsilon}(\omega), \hat{\mu}(\omega) \leftarrow \text{dispersion}$$

- Give a model for dispersion in dielectrics

## High Frequency & Linear Response: (Recap)

• Need to specify  $\vec{j}(t, \vec{x})$ :

• The most general linear form involving no spatial derivatives

$$\vec{j}(t, \vec{x}) = \int dt \sigma(t-t') \vec{E}(t', \vec{x}) + \text{spatial derivs suppressed by } \lambda_{\text{micro}}/L$$

• for a causal system  $\sigma(t-t')$  should have support only for  $t > t'$ , i.e.

$$\sigma(t) = 0 \quad \text{for } t < 0$$

•  $\vec{j}(t, \vec{x})$  is a convolution, fourier transforming

$$\vec{j}(\omega, \vec{x}) = \sigma(\omega) E(\omega, \vec{x})$$

↑  
frequency dependent conductivity

## Expectations for $\sigma(\omega)$ at low frequency

① For a conductor,  $\vec{j} = \sigma_0 E$

put  $\sigma_0$  to keep it apart  
from  $\sigma(t, t')$

Fourier transforming

$$j(t, x) = \sigma_0 E(t, x), \text{ we have}$$

$$j(\omega, x) = \sigma_0 E(\omega, x)$$

i.e.  $\sigma(\omega) = \sigma_0$  at low frequency

② For an insulator

$$\vec{j} = \partial_t \vec{P}$$

$$j(\omega, x) = -i\omega P$$

$$\approx -i\omega \chi_e E \quad \Leftrightarrow \quad \sigma(\omega) = -i\omega \chi_e$$

Thus we sometimes define for insulators

$$\sigma(\omega) = -i\omega \chi_e(\omega)$$

and  $\sigma(\omega) = -i\omega P(\omega)$

## Maxwell Eqs @ Dispersion

- Now we can continue and add the first derivative

$$\vec{j}(\omega) = -i\omega \chi_e(\omega) \vec{E}(\omega) + c \chi_m^B(\omega) \nabla \times \vec{B}(\omega, \vec{x})$$

- From the continuity equation, we have

$$\begin{aligned} -i\omega \rho(\omega) &= -\nabla \cdot \vec{j} \\ &= -i\omega \chi_e(\omega) (-\nabla \cdot \vec{E}) + \underbrace{\nabla \cdot \nabla \times}_{0} \end{aligned}$$

or,  $\rho(\omega) = \chi_e(\omega) (-\nabla \cdot \vec{E})$

- Thus the only difference between this and before is that now  $\chi_e(\omega)$  and  $\chi_m^B(\omega)$  are functions of frequency not constants

$$\epsilon(\omega) \nabla \cdot \vec{E} = 0$$

$$\nabla \times \vec{B} = \frac{\epsilon(\omega) \mu(\omega)}{c^2} (-i\omega \vec{E})$$

$$\nabla \cdot \vec{B} = 0$$

$$\nabla \times \vec{E} = + i\omega \frac{\vec{B}}{c}$$



Look for plane wave solutions

$$\vec{E}(x) = \vec{E}_0 e^{i\vec{k}\cdot\vec{x}}$$

Then:

$$\epsilon(\omega) \vec{k} \cdot \vec{E}_0 = 0 \iff \vec{E}_0 \text{ is transverse}$$

unless  $\epsilon(\omega(k)) = 0$   
(can happen)

$$i\vec{k} \times \vec{B}_0 = \frac{\epsilon\mu}{c^2} (-i\omega \vec{E}_0)$$

$$i\vec{k} \cdot \vec{B}_0 = 0$$

$$i\vec{k} \times \vec{E}_0 = \frac{\omega}{c} \vec{B}_0$$

We will ignore longitudinal modes, and consider only transverse modes  $\vec{E}_0 \cdot \vec{k} = 0$

$$\vec{k} \times (\vec{k} \times \vec{E}_0) = \frac{\omega}{c} \vec{k} \times \vec{B}_0$$

$$\vec{k} (\vec{k} \cdot \vec{E}_0) - \vec{k}^2 \vec{E}_0 = -\frac{\omega^2}{c^2} \epsilon(\omega) \mu(\omega) \vec{E}_0$$

0 for transverse modes

$$-k^2 + \frac{\omega^2}{c^2} \epsilon(\omega) \mu(\omega) = 0$$

Complex index of refraction

$$n^2(\omega) = \epsilon(\omega) \mu(\omega)$$

This determines  $\omega(\vec{k})$

## Propagation of Waves in dispersive media:

• Real part of  $\epsilon(\omega)$  determines the phase velocity (and group velocity)

• Im part of  $\epsilon(\omega)$  determines the absorption

To see this solve for the frequency, set  $\mu(\omega) = 1$

$$-k^2 + \frac{\omega^2}{c^2} \epsilon(\omega) = 0$$

And assume that the imaginary part is small

$$\epsilon(\omega) = \underbrace{\epsilon'(\omega)}_{\substack{\text{real} \\ \text{large}}} + i \underbrace{\epsilon''(\omega)}_{\text{im small}} \quad \omega = \omega_*(k) - i \underbrace{\Gamma(k)}_{\substack{2 \\ \text{small}}}$$

Then at zero order:

$$\boxed{-k^2 + \frac{\omega_*^2(k)}{c^2} \epsilon'(\omega_*(k)) = 0} \quad \Leftarrow \text{determines } \omega_*(k)$$

$$\omega_*(k) = \frac{ck}{\sqrt{\epsilon'(\omega_*)}} \equiv \boxed{\frac{ck}{n(\omega_*)} = \omega_*}$$

At first

$$\frac{c}{n(\omega_*)} = \frac{\omega_*}{k}$$

$$2\omega \left( \frac{i\Gamma}{2} \right) \epsilon' + i\omega^2 \epsilon''(\omega) = 0$$

Find using the zeroth order solution  $\omega \approx \omega_*(k)$

$$\Gamma(k) = \omega_* \frac{\epsilon''(\omega_*)}{\epsilon'(\omega_*)}$$

Thus the wave  $E = E_0 e^{-i\omega t} e^{ik \cdot x}$

$$E \approx E_0 e^{-i\omega_* t} e^{-\Gamma/2 t} e^{ik \cdot x}$$

## Simple model for $\sigma(\omega)$ for a dielectric:

Lets go back and revive the oscillator model

← Atoms electrons harmonically bound to protons

$$m \frac{d^2x}{dt^2} + m\gamma \frac{dx}{dt} + m\omega_0^2 = eE_\omega e^{-i\omega t}$$

Find  $x = x_\omega e^{-i\omega t}$

$$[-m\omega^2 - im\gamma\omega + m\omega_0^2] x_\omega = eE_\omega$$

So

$$x_\omega = \frac{(e/m) E_\omega}{-\omega^2 + \omega_0^2 - i\omega\gamma}$$

And  $j\omega = eN(-i\omega)x_\omega$       $j = eNv(t)$

$$j\omega = \frac{(Ne^2/m)}{-\omega^2 + \omega_0^2 - i\omega\gamma} (-i\omega E)$$

# Lorentz model for Dielectric

So



$$\chi_e(\omega) = \frac{(Ne^2/m)}{-\omega^2 + \omega_0^2 - i\omega\gamma}$$

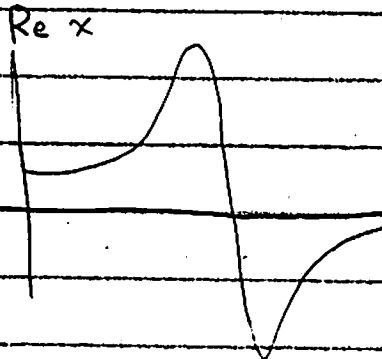
Find

$$\chi_e(\omega) = \text{Re } \chi_e + i \text{Im } \chi_e$$

$$\chi_e = \frac{(Ne^2/m)(\omega_0^2 - \omega^2)}{[(\omega^2 - \omega_0^2)^2 + (\omega\gamma)^2]}$$

$$+ \frac{(Ne^2/m) i\omega\gamma}{[(\omega^2 - \omega_0^2)^2 + (\omega\gamma)^2]}$$

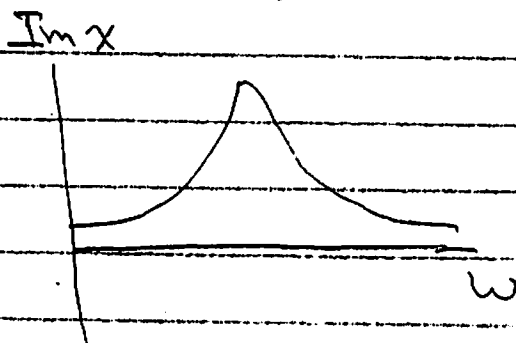
So



$$\epsilon \equiv 1 + \chi_e$$

$$\text{Re } \epsilon(\omega) = 1 + \text{Re } \chi_e$$

$$\text{Im } \epsilon(\omega) = \text{Im } \chi_e$$



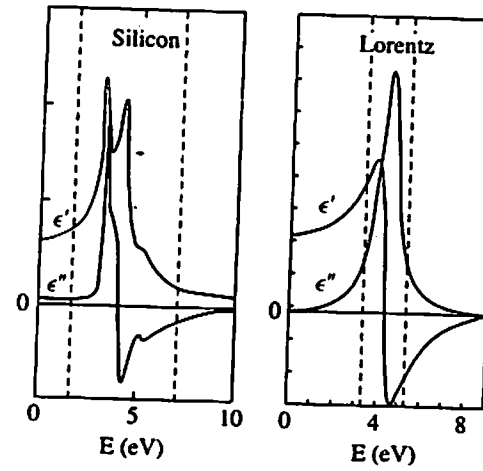


Figure 18.4: Real and imaginary parts of  $\tilde{\epsilon}(\omega)/\epsilon_0$  for silicon. Left panel: experiment. Right panel: Lorentz model. Vertical dashed lines are discussed in the text. Figure adapted from Wooten (1972).

## Last Time

- Discussed Propagation of waves in dispersive media

$$\vec{j}(t) = \int dt' \sigma(t-t') E(t') + \int dt' \chi_m^B(t-t') \nabla \times B(t')$$

$\sigma(t-t')$ , and  $\chi_m^B(t-t')$  are causal:

$$\chi_m^B(t) = 0 \quad \text{for } t < 0$$

$$\sigma(t) = 0 \quad \text{for } t < 0$$

- In Fourier space (in time)

$$\vec{j}(\omega) = \sigma(\omega) E(\omega, x) + \chi_m^B(\omega) \nabla \times B(\omega, x)$$

- After making these replacements, found that the Helmholtz equations for transverse waves reads

$$\left( \nabla^2 + \omega^2 \frac{\mu(\omega) \epsilon(\omega)}{c^2} \right) \vec{E}_T(\omega, x) = 0$$

- Found wave solutions: Define

$$\epsilon(\omega) \equiv 1 + \chi_e(\omega), \quad \sigma(\omega) \equiv -i\omega \chi_e(\omega)$$

$$\text{and } \mu(\omega) = 1/(1 - \chi_m^B(\omega))$$

Set  $\mu(\omega) = 1 / (1 - \chi_m^B(\omega)) = 1$ , then

$$-k^2 + \frac{\omega^2}{c^2} \epsilon(\omega(k)) = 0$$

Then solve for  $\omega(k) = \omega_*(k) - i\frac{\Gamma(k)}{2}t$ , so

$$\begin{aligned} E(t, x) &= E_0 e^{-i\omega(k)t} e^{ik \cdot x} \\ &\approx E_0 e^{-\Gamma/2 t} e^{-i\omega_*(k)t} e^{ik \cdot x} \end{aligned}$$

Where for small damping,  $\omega_*(k)$  found from

$$-k^2 + \frac{\omega_*^2}{c^2} \cdot \text{Re } \epsilon(\omega_*) = 0 \quad \leftarrow \text{this determines the dispersion curve } \omega_*(k)$$

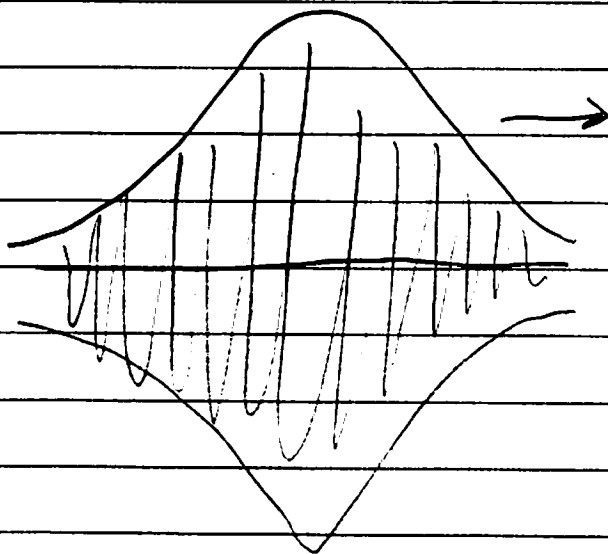
Then the imaginary part determines the damping rate

$$\Gamma = \omega \frac{\text{Im } \epsilon(\omega)}{\text{Re } \epsilon(\omega)}$$



## Wave Packets

- So far we have been considering individual plane waves. A general wave is a superposition of plane waves



The wave packet should also be a solution to the Helmholtz equations. This means for every  $\vec{k}$ , there is an  $\omega(\vec{k})$ . We will assume  $\omega(\vec{k})$  real. In general there is imaginary part. Then

$$u(x, t) = \int_{-\infty}^{\infty} \frac{dk}{2\pi} A(k) e^{ikx - i\omega(k)t} \sim \sum_k A_k e^{ik_n x - i\omega_k t}$$

The shape of the initial packet determines  $A(k)$

$$u(x, 0) = \int_{-\infty}^{\infty} \frac{dk}{2\pi} A(k) e^{ikx} \implies A(k) = \int_{-\infty}^{\infty} dx u(x, 0) e^{-ik \cdot x}$$

# Wave Packets Pg. 2

- Recall some fourier transforms

Gaussian:  $G(x) \equiv C e^{-x^2/4\sigma^2} \longleftrightarrow \hat{G}(k) = \tilde{C} e^{-k^2\sigma^2}$

phase:  $e^{ik_0 x} f(x) \longleftrightarrow \hat{f}(k - k_0)$   
vs.  $f(x - x_0) \longleftrightarrow e^{-ikx_0} \hat{f}(k)$   
shift

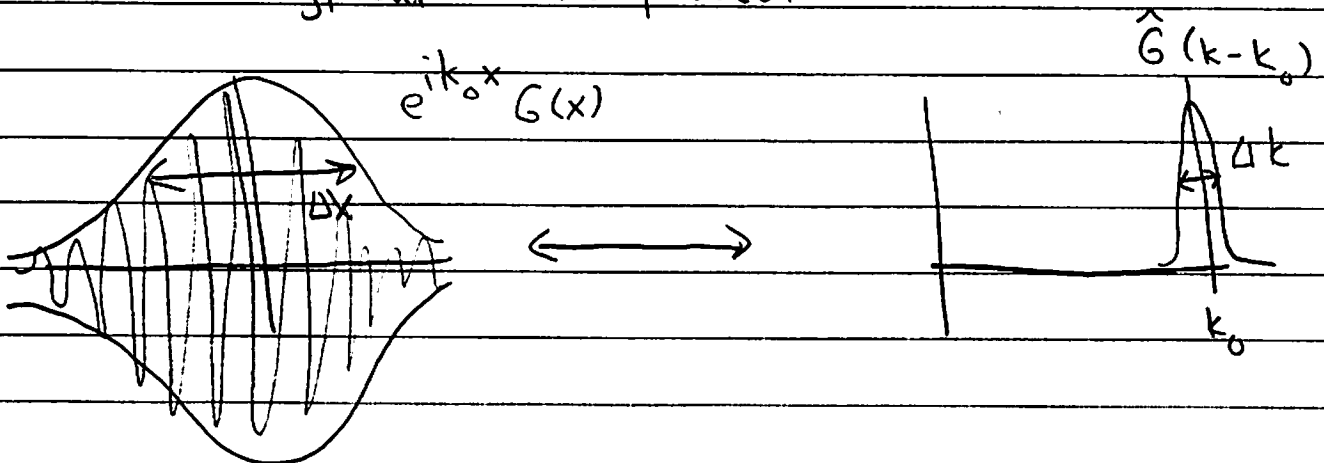
- The uncertainty principle applies to

$$(\Delta x)^2 = \int dx |u(x,0)|^2 (x - \bar{x})^2$$

$$(\Delta k)^2 = \int \frac{dk}{2\pi} |A(k)|^2 (k - \bar{k})^2$$

find  $\Delta k \Delta x \geq \frac{1}{2}$  with equality holding <sup>uniquely</sup> for gaussian

- So a typical wave packet



$$\Delta x \sim \frac{L}{\Delta k}$$

where  $\Delta k \ll k_0$  since  $k_0 \Delta x \sim \frac{k_0 L}{\Delta k} \gg 1$

- Then, let's ask about the solution at future times:

$$u(x, t) = \int_{-\infty}^{\infty} \frac{dk}{2\pi} A(k) e^{+ikx - i\omega(k)t}$$

And we expand near  $k_0$   $\left. \frac{d\omega}{dk} \right|_{k=k_0}$

$$\omega(k) \approx \omega(k_0) + \frac{d\omega_0}{dk} (k - k_0)$$

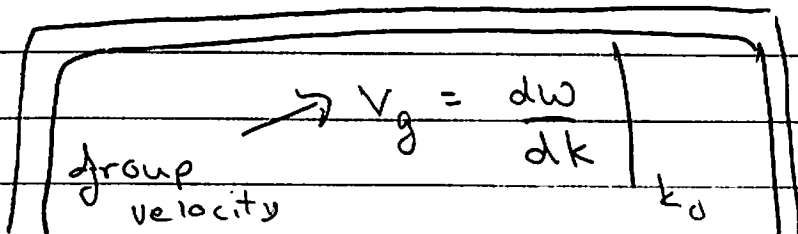
So

$$u(x, t) = \underbrace{e^{+i \left[ \frac{d\omega_0}{dk} k_0 - i\omega(k_0) \right] t}}_{e^{i\phi_0 t}} \int_{-\infty}^{\infty} \frac{dk}{2\pi} e^{ikx - \frac{d\omega_0}{dk} k t} A(k)$$

$$= e^{i\phi_0 t} \int_{-\infty}^{\infty} e^{ik(x - \frac{d\omega_0}{dk} t)} A(k)$$

$$u(x, t) = e^{i\phi_0 t} u(x - \frac{d\omega_0}{dk} t)$$

Thus we see that apart from an irrelevant phase, the wave packet travels with a speed given by



For

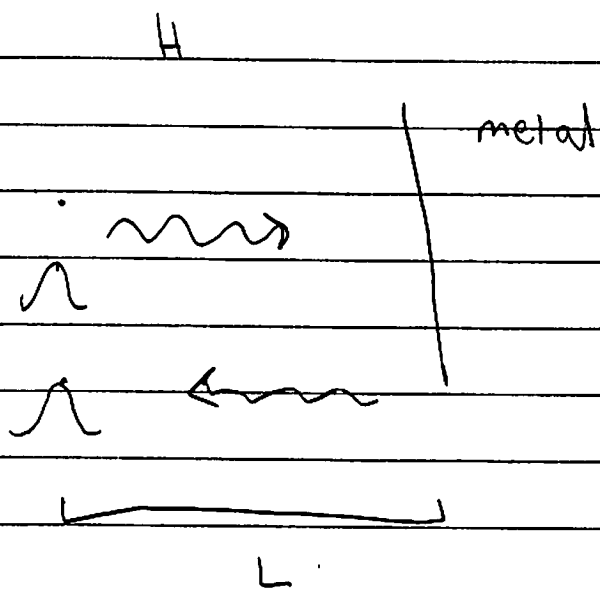
$$w(k) = \frac{ck}{n(k)}$$

$$\frac{dw}{dk} = \frac{c}{n(k)} - \frac{ck}{n^2} \frac{dn}{dw} \frac{dw}{dk}$$

Solve

$$\frac{dw}{dk} = \frac{c}{n(w) + dn/dw}$$

## Comments on Homework Problems:



Show that  $t_{\text{return}} = \frac{2L}{c} + \text{bit}$  time delay interacting with metal

How to calculate the time delay. We showed that

$$H_R(k) = H_I(k) \underbrace{r(k) e^{i\phi(k)}}_{\text{reflection amplitude}}$$

reflection coefficient  
 $R = |r e^{i\phi}|^2$

So

$$H_R(x, t) = \int \frac{dk}{2\pi} e^{-ikz - ckt} H_I(k) r(k) e^{i\phi(k)}$$

Now you can expand phase  $\phi$  (and the pre-amp  $r(k) \rightarrow R(k)$ ) to find  $H_R(x, t)$ , and see when the center of the wave packet returns to its starting point.

## Retarded Grn fcn's

- Take a damped harmonic oscillator

$$\underbrace{\left[ m \frac{d^2}{dt^2} + m\gamma \frac{d}{dt} + m\omega_0^2 \right]}_{\equiv \mathcal{L}_t} G_R(t, t_0) = \delta(t - t_0)$$

$G_R(t, t_0)$  is the displacement at time  $t$ , due to an impulsive force at time  $t_0$ . For a general  $F(t_0)$  driving the oscillator

$$x(t) = \int_{-\infty}^{\infty} dt_0 G_R(t - t_0) F(t_0) \leftarrow \begin{array}{l} \text{This the} \\ \text{inhomogeneous} \\ \text{solution. Later} \\ \text{we will add the} \\ \text{homogeneous sol.} \end{array}$$

Since

$$\begin{aligned} \mathcal{L}_t x(t) &= \int_{-\infty}^{\infty} dt_0 \mathcal{L}_t G_R(t - t_0) F(t_0) \\ &= \int_{-\infty}^{\infty} dt_0 \delta(t - t_0) F(t_0) = F(t) \end{aligned}$$

- my main goal is to write down the retarded green-fcn of the maxwell eqs

- Demand causality  $G_R(t) = 0$  for  $t < 0$ ,  
i.e.

$$G_R(t, t_0) = 0 \quad \text{for } t < t_0$$